

GENERAL DESCRIPTION

OB2353 combines a dedicated current mode PWM controller with integrated high voltage power MOSFET .It is optimized for high performance low standby power and cost effective offline flyback converter applications in sub 10W range.

OB2353 PWM frequency operates in the fixed and trimmed. At no load or light load the frequency is internal adjusted, the IC operates in Extended 'burst mode' to minimize switching loss. Lower standby power and higher conversion efficiency is thus achieved.

VDD low startup current and low operating current contribute to a reliable power on startup design with OB2353. A large value resistor could thus be used in the startup circuit to minimize the standby power.

The internal slope compensation improves system large signal stability and reduces the possible sub-harmonic oscillation at high PWM duty cycle output. Leading-edge blanking on current sense(CS) input removes the signal glitch due to snubber circuit diode reverse recovery and thus greatly reduces the external component count and system cost in the design.

OB2353 offers complete protection coverage with automatic self-recovery feature including Cycle-by-Cycle current limiting (OCP), over load protection (OLP), VDD over voltage clamp and under voltage lockout (UVLO). Excellent EMI performance is achieved with On-Bright proprietary frequency shuffling technique together with soft switching control at the totem pole gate drive output.

The tone energy at below 20KHZ is minimized in the design and audio noise is eliminated during operation. OB2353 is offered in SOP-8 package.

FEATURES

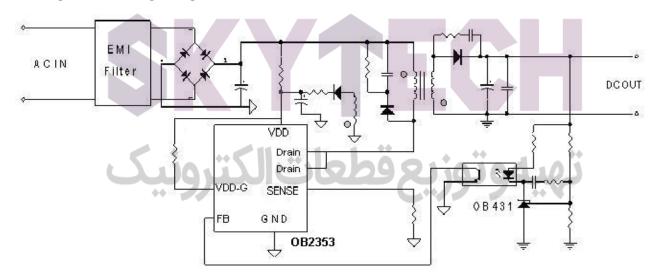
- Optimized for sub 10W Applications
- Single Chip 650V Sense FET Power Switch for Package.
- Soft Start with 4mS
- Frequency shuffling for EMI
- Extended Burst Mode Control For Improved Efficiency and Minimum Standby Power Design
- Audio Noise Free Operation
- Fixed 50KHZ Switching Frequency
- Internal Synchronized Slope Compensation
- Low VDD Startup Current and Low Operating Current
- Leading Edge Blanking on Current Sense Input
- Good Protection Coverage With Auto Self-Recovery
 - VDD Over Voltage Clamp and Under Voltage Lockout with Hysteresis (UVLO)
 - On-Bright Proprietary Line Input Compensated Cycle-by-Cycle Over-current Threshold Setting For Constant Output Power Limiting Over Universal Input Voltage Range.
 - Overload Protection (OLP).
 - o Over voltage Protection(OVP)

APPLICATIONS

Offline AC/DC flyback converter for

- Battery Charger
- PDA power supplies
- Digital Cameras
- Power Adaptor
- Set-Top Box Power Supplies
- Open-frame SMPS

TYPICAL APPLICATION

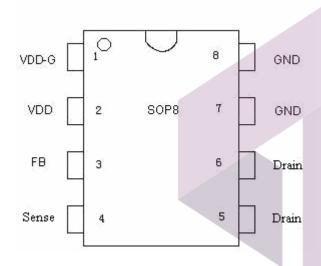




GENERAL INFORMATION

Pin Configuration

The OB2353 is offered in SOP8 package, shown as below.



Part Number	Description
OB2353CP	SOP8, Pb-free
OB2353CP/A	SOP8,Pb-free,Tape&Reel

Package Dissipation Rating

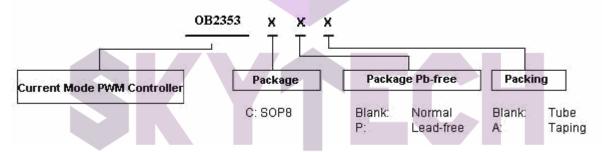
I delinge Dissipation It	******
	RθJA (°C/W)
SOP8	150

Absolute Maximum Ratings

Parameter	Value
Drain Input Voltage	650V
VDD Input Voltage	32 V
VDD-G Input Voltage	32 V
VDD Input DC Current	10mA
V _{FB} Input Voltage	-0.3 to 7V
V _{SENSE} Input Voltage to Sense	-0.3 to 7V
Pin	
Min/Max Operating Junction	-20 to 150 °C
Temperature T _J	
Min/Max Storage Temperature	-55 to 160 °C
T_{stg}	
Lead Temperature (Soldering,	260 °C
10secs)	

Note: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.

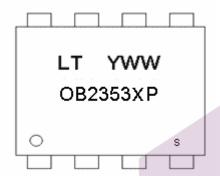
Ordering Information







Marking Information



X: C for SOP8;

P: Pb-free Package;

Y: Year Code (0-9)

WW: Week Code (1-52) s: Optional Internal Code

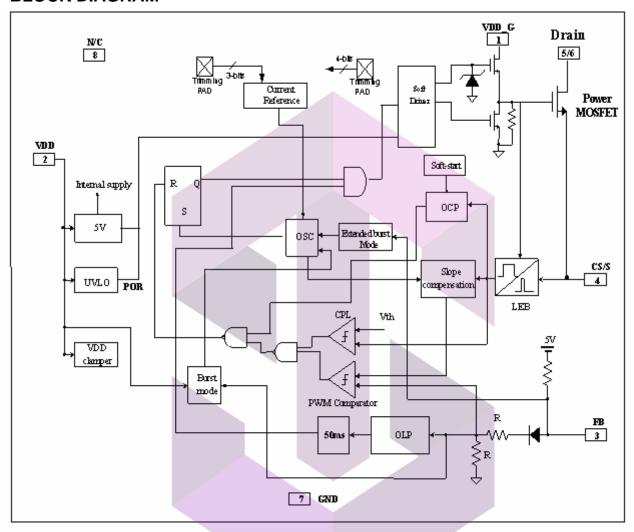
TERMINAL ASSIGNMENTS

Pin Name	I/O	Description
GND	P	Ground
FB	I	Feedback input pin. The PWM duty cycle is determined by voltage level into this pin and
		the current-sense signal at Pin 3.
VDD-G	P	Power Supply specially for Internal Gate Driver
SENSE	I	Current sense input pin.
VDD	P	Chip DC power supply pin.
Drain	P	The Drain pin is designed to connect directly to the primary lead of the transformer and
		is capable of switching a maximum of 650V





BLOCK DIAGRAM



SKYTECH تهيه وتوزيع قطعات الكترونيك



ELECTRICAL CHARACTERISTICS

 $(T_A = 25^{\circ}C \text{ if not otherwise noted})$

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
Supply Voltage (VI	OD)					
I_VDD_Startup	VDD Start up	VDD=14.5V,Measure		5	20	uA
	Current	Leakage current into VDD				
I_VDD_Operation	Operation Current	$VDD=16V, V_{FB}=3V$		1.6		mA
UVLO(ON)	VDD Under		8.7	9.7	10.7	V
	Voltage Lockout					
	Enter					
UVLO(OFF)	VDD Under		14.9	15.9	16.9	V
	Voltage Lockout					
	Exit (Recovery)					
OVP(ON)	Over voltage	VDD=16V,CS=0V,FB=3V	28.5	30	31.5	V
	protection	Ramp up VDD until gate				
	voltage	clock is off				
VDD Clamp	VDD Zener Clamp	$I_{DD} = 10 \text{ mA}$	31	32	33	V
_ r	Voltage					
Feedback Input Sec					1	1
V _{FB} Open	V _{FB} Open Loop		5.4	5.7	6.0	V
12_ 1	Voltage					
I _{FB} _Short	FB pin short	Short FB pin to GND and		1.45		mA
	circuit current	measure current				
V_{TH}_0D	Zero Duty Cycle	VDD=16V		0.8		V
_	FB Threshold					
	Voltage					
V _{TH} _PL	Power Limiting			3.7		V
_	FB Threshold					
	Voltage					
T_{D} PL	Power limiting			50		mSec
	Debounce Time					
Z_{FB}_{IN}	Input Impedance			4		Kohm
Current Sense Inpu	ıt(Sense Pin)					
Soft start time				4		ms
T_blanking	Leading edge			270		ns
	blanking time					
Z _{SENSE} _IN	Input Impedance			40		Kohm
T _D OC	Over Current	From Over Current Occurs		120		nSec
	Detection and	till the Gatedrive output start				
	Control Delay	to turn off				
V _{TH} _OC	Internal Current	FB=3.3V	0.72	0.77	0.82	V
	Limiting	VDD=16V				
	Threshold Voltage					
Oscillator						
		lana lana lana	- ee		1 200	T
Fosc	Normal Oscillation Frequency	VDD = 16V	45	50	55	KHZ
△f Temp	Frequency	VDD = 16V		5	Ų.	%
	Temperature					
Δ1_1emp	Temperature					
	Stability					
$\triangle f_VDD$		VDD = 16-20V		5		%
_ ^	Stability	VDD = 16-20V		5		%



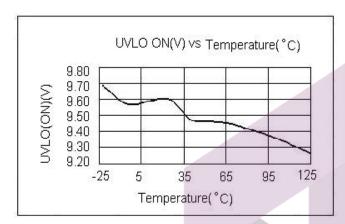
D max	Maximum duty	FB=3.3V, CS =0V	70	80	90	%	
	cycle	VDD=16V					
F_Burst	Burst Mode Base	VDD = 16V		22		KHZ	
	Frequency						
Mosfet Section	Mosfet Section						
Vds	Drain-Source		650			V	
	Breakdown						
	Voltage						
RDS(on)	Static Drain-			12	15	Ω	
	Source On-						
	Resistance						
Idp	Drain current-peak				0.9	A	
Frequency Shuffling							
Δf_{OSC}	Frequency		-4		4	%	
	Modulation range						
	/Base frequency						

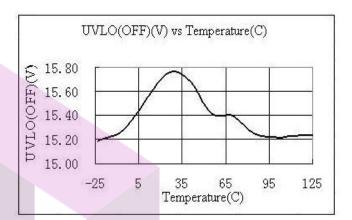


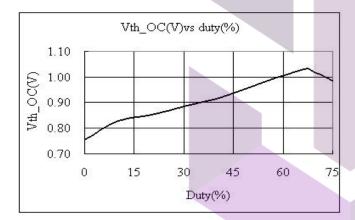


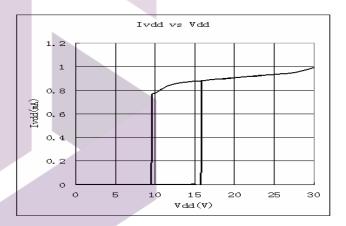
CHARACTERIZATION PLOTS

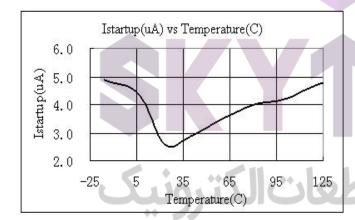
(The characteristic graphs are normalized at Ta=25 $^{\circ}$ C)

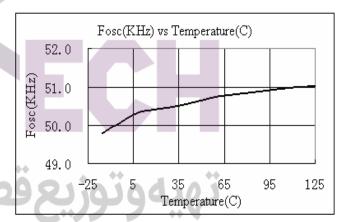














OPERATION DESCRIPTION

The OB2353 is a Low Power Off Line SMPS Switcher optimized for offline flyback converter applications in sub 10W power range. The extended burst mode control greatly reduces the standby power consumption and helps the design easily to meet the international power conservation requirements.

Startup Current and Start up Control

Startup current of OB2353 is designed to be very low so that VDD could be charged up above UVLO threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet reliable startup in application. For AC/DC adaptor with universal input range design, a 2 M Ω , 1/8 W startup resistor could be used together with a VDD capacitor to provide a fast startup and yet low power dissipation design solution.

• Operating Current

The Operating current of OB2353 is low at 2mA. Good efficiency is achieved with OB2353 low operating current together with extended burst mode control features.

• Frequency shuffling for EMI improvement

The frequency Shuffling (switching frequency modulation) is implemented in OB2353. The oscillation frequency is modulated with a random source so that the tone energy is spread out. The spread spectrum minimizes the conduction band EMI and therefore eases the system design.

• Extended Burst Mode Operation

At light load or zero load condition, most of the power dissipation in a switching mode power supply is from switching loss on the mosfet transistor, the core loss of the transformer and the loss on the snubber circuit. The magnitude of power loss is in proportion to the switching frequency. Lower switching frequency leads to the reduction on the power loss and thus conserves the energy.

At no load or light load the frequency is internal adjusted. The switch frequency reduces at light/no load condition to improve the conversion efficiency. At light load or no load condition, the FB input drops below burst mode threshold level and device enters Burst Mode control. The Gate drive output switches only when VDD voltage drops below a preset level and FB input is active to output an on

state. Otherwise the gate drive remains at off state to minimize the switching loss and reduces the standby power consumption to the greatest extend. The frequency control also eliminates the audio noise at any loading conditions.

• Oscillator Operation

The switching frequency of OB2353 is internally fixed at 50KHZ. No external frequency setting resistor is required for PCB design simplification.

• Current Sensing and Leading Edge Blanking
Cycle-by-Cycle current limiting is offered in
OB2353 current mode PWM control. The switch
current is detected by a sense resistor into the sense
pin. An internal leading edge blanking circuit chops
off the sense voltage spike at initial internal mosfet
on state due to snubber diode reverse recovery and
surge gate current of internal Mosfet so that the
external RC filtering on sense input is no longer
needed. The current limit comparator is disabled
and cannot turn off the internal Mosfet during the
blanking period. The PWM duty cycle is
determined by the current sense input voltage and
the FB input voltage.

• Internal Synchronized Slope Compensation

Built-in slope compensation circuit adds voltage ramp onto the current sense input voltage for PWM generation. This greatly improves the close loop stability at CCM and prevents the sub-harmonic oscillation and thus reduces the output ripple voltage.

• Drive

OB2353 driver internal Mosfet gate for power switch control. Too weak the gate drive strength results in higher conduction and switch loss of mosfet while too strong gate drive results the compromise of EMI

A good tradeoff is achieved through the built-in totem pole gate design with right output strength and dead time control. The low idle loss and good EMI system design is easier to achieve with this dedicated control scheme. An internal 16V clamp is added for internal Mosfet gate protection at higher than expected VDD input.



• Protection Controls

Good power supply system reliability is achieved with its rich protection features including Cycle-by-Cycle current limiting (OCP), Over Load Protection (OLP) and over voltage clamp, Under Voltage Lockout on VDD (UVLO).

With On-Bright Proprietary technology, the OCP is line voltage compensated to achieve constant output power limit over the universal input voltage range.

At overload condition when FB input voltage exceeds power limit threshold value for more than TD_PL, control circuit reacts to shut down the internal power Mosfet. Device restarts when VDD voltage drops below UVLO limit.

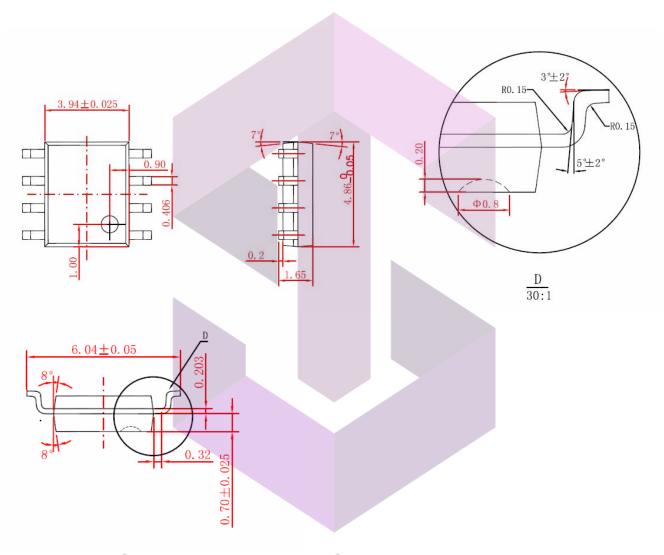
VDD is supplied by transformer auxiliary winding output. It is clamped when VDD is higher than 30V. The internal mosfet is shut down when VDD drops below 9.7V UVLO limit and device enters power on start-up sequence thereafter.





PACKAGE MECHANICAL DATA

8-Pin Plastic SOP



SKYTECH Sie "Slewlebäere"



IMPORTANT NOTICE

RIGHT TO MAKE CHANGES

On-Bright Electronics Corp. reserves the right to make corrections, modifications, enhancements, improvements, and other changes to its products and services at any time and to discontinue any product or service without notice. Customers should obtain the latest relevant information before placing orders and should verify that such information is current and complete.

WARRANTY INFORMATION

On-Bright Electronics Corp. warrants performance of its hardware products to the specifications applicable at the time of sale in accordance with its standard warranty. Testing and other quality control techniques are used to the extent it deems necessary to support this warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily performed.

On-Bright Electronics Corp. assumes no liability for applications assistance or customer product design. Customers are responsible for their products and applications using On-Bright's components, data sheet and application notes. To minimize the risks associated with customer products and applications, customers should provide adequate design and operating safeguards.

LIFE SUPPORT

On-Bright Electronics Corp.'s products are not designed to be used as components in devices intended to support or sustain human life. On-bright Electronics Corp. will not be held liable for any damages or claims resulting from the use of its products in medical applications.

On-Bright Electronics Corp.'s products are not designed for use in military applications. On-Bright Electronics Corp. will not be held liable for any damages or claims resulting from the use of its products in military applications.

